# $\frac{\text{UNIT 8}}{\text{Z-Transforms} - 2}$

# 8.1 Transform analysis of LTI systems:

 We have defined the transfer function as the z-transform of the impulse response of an LTI system

$$H(z) = \sum_{k=-\infty}^{\infty} h[k]z^{-k}$$

- Then we have y[n] = x[n] \* h[n] and Y(z) = X(z)H(z)
- This is another method of representing the system
- The transfer function can be written as

$$H(z) = \frac{Y(z)}{X(z)}$$

- This is true for all z in the ROCs of X(z) and Y(z) for which X(z) in nonzero
- The impulse response is the *z*-transform of the transfer function
- We need to know ROC in order to uniquely find the impulse response
- If ROC is unknown, then we must know other characteristics such as stability or causality in order to uniquely find the impulse response

### System identification

- Finding a system description by using input and output is known as system identification
- Ex1: find the system, if the input is  $x[n] = (-1//3)^n u[n]$  and the out is  $y[n] = 3(-1)^n u[n] + (1/3)^n u[n]$



• Solution: Find the *z*-transform of input and output. Use X(z) and Y(z) to find H(z), then find h(n) using the inverse *z*-transform

$$X(z) = \frac{1}{(1 + (\frac{1}{3})z^{-1})},$$
 with ROC  $|z| > \frac{1}{3}$ 

$$Y(z) = \frac{3}{(1+z^{-1})} + \frac{1}{(1-(\frac{1}{3})z^{-1})}, \text{ with ROC } |z| > 1$$

• We can write Y(z) as

$$Y(z) = \frac{4}{(1+z^{-1})(1-(\frac{1}{3})z^{-1})}, \text{ with ROC } |z| > 1$$

• We know H(z) = Y(z)/X(z), so we get

$$H(z) = \frac{4(1+(\frac{1}{3})z^{-1})}{(1+z^{-1})(1-(\frac{1}{3})z^{-1})} \quad \text{with ROC} \quad |z| > 1$$

• We need to find inverse *z*-transform to find x[n], so use partial fraction and write H(z) as

$$H(z) = \frac{2}{1+z^{-1}} + \frac{2}{1-(\frac{1}{3})z^{-1}}$$
 with ROC  $|z| > 1$ 

• Impulse response x[n] is given by

$$h[n] = 2(-1)^n u[n] + 2(1/3)^n u[n]$$

## Relation between transfer function and difference equation

- The transfer can be obtained directly from the difference-equation description of an LTI system
- We know that

$$\sum_{k=0}^{N} a_k y[n-k] = \sum_{k=0}^{M} b_k x[n-k]$$

- We know that the transfer function H(z) is an eigen value of the system associated with the eigen function  $z^n$ , ie. if  $x[n] = z^n$  then the output of an LTI system  $y[n] = z^n H(z)$
- Put  $x[n-k] = z^{n-k}$  and  $y[n-k] = z^{n-k}H(z)$  in the difference equation,



we get

$$z^{n} \sum_{k=0}^{N} a_{k} z^{-k} H(z) = z^{n} \sum_{k=0}^{M} b_{k} z^{-k}$$

• We can solve for H(z)

$$H(z) = \frac{\sum_{k=0}^{M} b_k z^{-k}}{\sum_{k=0}^{N} a_k z^{-k}}$$

- The transfer function described by a difference equation is a ratio of polynomials in  $z^{-1}$  and is termed as a rational transfer function.
- The coefficient of  $z^{-k}$  in the numerator polynomial is the coefficient associated with x[n-k] in the difference equation
- The coefficient of  $z^{-k}$  in the denominator polynomial is the coefficient associated with y[n-k] in the difference equation
- This relation allows us to find the transfer function and also find the difference equation description for a system, given a rational function



### **Transfer function:**

- The poles and zeros of a rational function offer much insight into LTI system characteristics
- The transfer function can be expressed in pole-zero form by factoring the numerator and denominator polynomial
- If  $c_k$  and  $d_k$  are zeros and poles of the system respectively and  $\tilde{b} = b_0/a_0$  is the gain factor, then

$$H(z) = \frac{\tilde{b} \prod_{k=1}^{M} (1 - c_k z^{-1})}{\prod_{k=1}^{N} (1 - d_k z^{-1})}$$

- This form assumes there are no poles and zeros at z = 0
- The  $p^{th}$  order pole at z = 0 occurs when  $b_0 = b_1 = \ldots = b_{p-1} = 0$
- The  $I^{th}$  order zero at z = 0 occurs when  $a_0 = a_1 = \ldots = a_{l-1} = 0$
- Then we can write H(z) as

$$H(z) = \frac{\tilde{b}z^{-p} \prod_{k=1}^{M-p} (1 - c_k z^{-1})}{z^{-l} \prod_{k=1}^{N-l} (1 - d_k z^{-1})}$$

where  $\tilde{b} = b_p/a_l$ 

- In the example we had first order pole at z = 0
- ullet The poles, zeros and gain factor  $\tilde{b}$  uniquely determine the transfer function
- This is another description for input-output behavior of the system
- The poles are the roots of characteristic equation



### 8.2 <u>Unilateral Z- transforms:</u>

- · Useful in case of causal signals and LTI systems
- The choice of time origin is arbitrary, so we may choose n = 0 as the time at which the input is applied and then study the response for times n > 0

## Advantages

- We do not need to use ROCs
- It allows the study of LTI systems described by the difference equation with initial conditions

### Unilateral z-transform

• The unilateral z-transform of a signal x[n] is defined as

$$X(z) = \sum_{n=0}^{\infty} x[n]z^{-n}$$

which depends only on x[n] for  $n \ge 0$ 

 The unilateral and bilateral z-transforms are equivalent for causal signals

$$\alpha^{n}u[n] \stackrel{z_{u}}{\longleftrightarrow} \frac{1}{1 - \alpha z^{-1}}$$

$$a^{n}\cos(\Omega_{o}n)u[n] \stackrel{z_{u}}{\longleftrightarrow} \frac{1 - a\cos(\Omega_{o})z^{-1}}{1 - 2a\cos(\Omega_{o})z^{-1} + a^{2}z^{-2}}$$

# Properties of unilateral Z transform:

Consider the difference equation description of an LTI system

$$\sum_{k=0}^{N} a_k y[n-k] = \sum_{k=0}^{M} b_k x[n-k]$$

• We may write the z-transform as

$$A(z)Y(z) + C(z) = B(z)X(z)$$

where

$$A(z) = \sum_{k=0}^{N} a_k z^{-k}$$
 and  $B(z) = \sum_{k=0}^{M} b_k z^{-k}$ 



- The same properties are satisfied by both unilateral and bilateral *z*-transforms with one exception: the time shift property
- The time shift property for unilateral *z*-transform: Let w[n] = x[n-1]
  - The unilateral z-transform of w[n] is

$$W(z) = \sum_{n=0}^{\infty} w[n] z^{-n} = \sum_{n=0}^{\infty} x[n-1] z^{-n}$$

$$W(z) = x[-1] + \sum_{n=1}^{\infty} x[n-1] z^{-n}$$

$$W(z) = x[-1] + \sum_{m=0}^{\infty} x[m] z^{-(m+1)}$$

• The unilateral z-transform of w[n] is

$$W(z) = x[-1] + z^{-1} \sum_{m=0}^{\infty} x[m] z^{-m}$$
 
$$W(z) = x[-1] + z^{-1} X(z)$$

- A one-unit time shift results in multiplication by  $z^{-1}$  and addition of the constant x[-1]
- In a similar way, the time-shift property for delays greater than unity is

$$x[n-k] \stackrel{z_u}{\longleftrightarrow} x[-k] + x[-k+1]z^{-1} + \dots + x[-1]z^{-k+1} + z^{-k}X(z) \text{ for } k > 0$$

• In the case of time advance, the time-shift property changes to

$$x[n+k] \stackrel{z_u}{\longleftrightarrow} -x[0]z^k - x[-1]z^{k-1} +$$

$$\dots -x[k-1]z + z^k X(z) \text{ for } k > 0$$



# 8.3 <u>Application to solve difference equations</u>

# Solving Differential equations using initial conditions:

We get

$$C(z) = \sum_{m=0}^{N-1} \sum_{k=m+1}^{N} a_k y[-k+m] z^{-m}$$

• We have assumed that x[n] is causal and

$$x[n-k] \stackrel{Z_u}{\longleftrightarrow} z^{-k}X(z)$$

- The term C(z) depends on the N initial conditions  $y[-1], y[-2], \dots, y[-N]$  and the  $a_k$
- C(z) is zero if all the initial conditions are zero
- Solving for Y(z), gives

$$Y(z) = \frac{B(z)}{A(z)}X(z) - \frac{C(z)}{A(z)}$$

- The output is the sum of the forced response due to the input and the natural response induced by the initial conditions
- The forced response due to the input

$$\frac{B(z)}{A(z)}X(z)$$

• The natural response induced by the initial conditions

$$\frac{C(z)}{A(z)}$$

• C(z) is the polynomial, the poles of the natural response are the roots of A(z), which are also the poles of the transfer function



• The form of natural response depends only on the poles of the system, which are the roots of the characteristic equation

### First order recursive system

• Consider the first order system described by a difference equation

$$y[n] - \rho y[n-1] = x[n]$$

where  $\rho = 1 + r/100$ , and r is the interest rate per period in percent and y[n] is the balance after the deposit or withdrawal of x[n]

•

- Assume bank account has an initial balance of \$10,000/- and earns 6% interest compounded monthly. Starting in the first month of the second year, the owner withdraws \$100 per month from the account at the beginning of each month. Determine the balance at the start of each month.
- Solution: Take unilateral z-transform and use time-shift property we get

$$Y(z) - \rho(y[-1] + z^{-1}Y(z)) = X(z)$$

• Rearrange the terms to find Y(z), we get

$$(1 - \rho z^{-1})Y(z) = X(z) + \rho y[-1]$$

$$Y(z) = \frac{X(z)}{1 - \rho z^{-1}} + \frac{\rho y[-1]}{1 - \rho z^{-1}}$$



- Y(z) consists of two terms
  - one that depends on the input: the forced response of the system
  - another that depends on the initial conditions: the natural response of the system
- The initial balance of \$10,000 at the start of the first month is the initial condition y[-1], and there is an offset of two between the time index n and the month index
- y[n] represents the balance in the account at the start of the  $n + 2^{nd}$  month.
- We have  $\rho = 1 + \frac{\frac{6}{12}}{100} = 1.005$
- Since the owner withdraws \$100 per month at the start of month 13 (n = 11)
- We may express the input to the system as x[n] = -100u[n-11], we get

$$X(z) = \frac{-100z^{-11}}{1 - z^{-1}}$$

• We get

$$Y(z) = \frac{-100z^{-11}}{(1-z^{-1})(1-1.005z^{-1})} + \frac{1.005(10,000)}{1-1.005z^{-1}}$$

· After a partial fraction expansion we get

$$Y(z) = \frac{20,000z^{-11}}{1 - z^{-1}} + \frac{20,000z^{-11}}{1 - 1.005z^{-1}} + \frac{10,050}{1 - 1.005z^{-1}}$$

ullet Monthly account balance is obtained by inverse z-transforming Y(z) We get

$$y[n] = 20,000u[n-11] - 20,000(1.005)^{n-11}u[n-11]$$
  
+  $10,050(1.005)^nu[n]$ 

- The last term  $10,050(1.005)^n u[n]$  is the natural response with the initial balance
- The account balance
- · The natural balance
- The forced response



# **Recommended Questions**

1. Find the inverse Z transform of 
$$H(z) = \frac{1+Z}{(1-0.9\,e^{j\pi/4}z^{-1})(1-0.9\,e^{-j\pi/4}z^{-1})}$$

2. A system is described by the difference equation

$$Y(n) - yn - 1) + \frac{1}{4}y(n - 2) = x(n) + 1/4x(n - 1) - 1/8x(n - 2)$$

Find the Transfer function of the Inverse system

Does a stable and causal Inverse system exists

- 3. Sketch the magnitude response for the system having transfer functions.
- 4. Find the z-transform of the following x[n]:

(a) 
$$x[n] = \{\frac{1}{2}, 1, -\frac{1}{3}\}$$

(b) 
$$x[n] = 2\delta[n+2] - 3\delta[n-2]$$

(c) 
$$x[n] = 3(-\frac{1}{2})^n u[n] - 2(3)^n u[-n-1]$$

(d) 
$$x[n] = 3(\frac{1}{2})^n u[n] - 2(\frac{1}{4})^n u[-n-1]$$

5. Given

$$X(z) = \frac{z(z-4)}{(z-1)(z-2)(z-3)}$$

- (a) State all the possible regions of convergence.
- (b) For which ROC is X (z) the z-transform of a causal sequence?
- 6. Show the following properties for the z-transform.

(a) If 
$$x[n]$$
 is even, then  $X(z^{-1}) = X(z)$ .

(b) If 
$$x[n]$$
 is odd, then  $X(z^{-1}) = -X(z)$ .

- (c) If x[n] is odd, then there is a zero in X(z) at z = 1.

7. Derive the following transform pairs: 
$$(\cos \Omega_0 n) u[n] \longleftrightarrow \frac{z^2 - (\cos \Omega_0) z}{z^2 - (2\cos \Omega_0) z + 1} \qquad |z| > 1$$

$$(\sin \Omega_0 n)u[n] \longleftrightarrow \frac{(\sin \Omega_0)z}{z^2 - (2\cos \Omega_0)z + 1}$$
  $|z| > 1$ 

8. Find the z-transforms of the following x[n]:

(a) 
$$x[n] = (n-3)u[n-3]$$

(b) 
$$x[n] = (n-3)u[n]$$

(c) 
$$x[n] = u[n] - u[n-3]$$

(d) 
$$x[n] = n\{u[n] - u[n-3]\}$$



9. Using the relation

$$a^n u[n] \longleftrightarrow \frac{z}{z-a}$$
  $|z| > |a|$ 

find the z-transform of the following **x[n]**:

(a) 
$$x[n] = na^{n-1}u[n]$$

(b) 
$$x[n] = n(n-1)a^{n-2}u[n]$$

(c) 
$$x[n] = n(n-1) \cdot \cdot \cdot (n-k+1)a^{n-k}u[n]$$

10. Using the z-transform

(a) 
$$x[n] * \delta[n] = x[n]$$

(b) 
$$x[n] * \delta[n - n_0] = x[n - n_0]$$

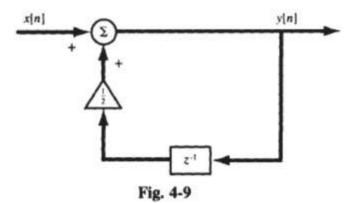
- 11. Find the inverse z-transform of  $X(z)=e^{a/z}$ , z>0
- 12. Using the method of long division, find the inverse z-transform of the following **X(z)**:

(a) 
$$X(z) = \frac{z}{(z-1)(z-2)}, |z| < 1$$

(b) 
$$X(z) = \frac{z}{(z-1)(z-2)}, 1 < |z| < 2$$

(c) 
$$X(z) = \frac{z}{(z-1)(z-2)}, |z| > 2$$

13. Consider the system shown in Fig. 4-9. Find the system function H(z) and its impulse response h[n]



14. Consider a discrete-time LTI system whose system function H(z) is given by

$$H(z) = \frac{z}{z - \frac{1}{2}}$$
  $|z| > \frac{1}{2}$ 

- (a) Find the step response s[n].
- (b) Find the output y[n] to the input x[n] = nu[n].
- 15. Consider a causal discrete-time system whose output y[n] and input x[n] are related by

$$y[n] - \frac{5}{6}y[n-1] + \frac{1}{6}y[n-2] = x[n]$$

- (a) Find its system function H(z).
  - (b) Find its impulse response h[n].

